



System Level Considerations with GaN Power Switching

Preface March 7<sup>th</sup>, 2018

APEC Industry Session - San Antonio, Texas

Authors: Peter Di Maso, GaN Systems Inc.; Di Chen, GaN Systems Inc.

### Digest

Gallium Nitride (GaN) transistors have been implemented into many power electronics applications by innovative and mainstream power electronics companies. GaN E-HEMTs (high electron mobility transistors) are used in applications for wireless charging and power transfer, DC motors for appliances, data center power, factory automation, solar, energy storage systems, and electric and autonomous vehicles. GaN enables the significant increase in efficiency that reduces losses, diminishes size, increases power density, prolongs battery life, and reduces bill-of-materials, capital expenditure and operating costs. All desirable outcomes in the mentioned applications. This presentation will demonstrate how the unique characteristics of GaN enable the performance of the HEMTs to produce the energy, weight, size and system cost reductions.

This presentation delve into the key parameters that affect performance, like gate charge, capacitance, drain-to-source resistance (including dynamic effects) and transconductance. The effects of temperature on these parameters will be discussed.

Topologies with that benefit from GaN E-HEMTs will be analyzed. The topologies to be analyzed include bridgeless totem pole for PFC, LLC half-bridges, phase-shifted full bridge. Very low dead times and fast switching are being enabled by GaN's low latency, fast switching, high frequency, ease of driving and high current capability. The low dead times further enable innovation in control techniques, new topologies and magnetics.

Thermal characteristics of GaN and its packaging will be demonstrated for enhancing system level thermals and packaging effectiveness. GaN qualification, reliability and standards have been popular topics in the power electronics industry. A discussion on the topic will demonstrate the how the wide bandgap transistor industry is coming together create standards and guidelines for test methods, reliability, qualification procedures, and datasheet parametrics for GaN based power conversion devices.

The presentation is aimed at intermediate power electronic design engineers and system engineers. The benefits of GaN transistors will also be of interest to power electronics technical marketing, business development, quality, reliability and component engineering professionals.

## Biographies



**Peter Di Maso** is director of product line management at GaN Systems and is responsible for creating and executing a sustainable portfolio of GaN power products. Peter has more than 20 years' experience in the power electronics industry. Peter's background includes strategic product marketing at Texas Instruments, where he helped establish TI's LED lighting product line. Peter also has gained experience from, and contributed to the success of leading power management semiconductor companies Maxim Integrated Products, Analog Devices and Allegro Microsystems.

Peter Di Maso has a Bachelor of Engineering degree (Electrical) from Concordia University in Montreal, QC, Canada and a Masters of Business Administration from Southern New Hampshire University in Manchester, NH, USA



Di Chen received his M.Sc in power electronics from Univ. of British Columbia in 2009 with research interest in power electronics design for renewable energy hybrid power systems. He has 8 years of power electronic system design experience.

In 2011, Di joined Honeywell and was responsible for power inverter design for industrial process control. He joined GaN Systems in 2014 where he has been working on customer application support and reference designs for next generation Gallium Nitride (GaN) HEMT based power systems.

Di Chen is in responsible for global application support, manages the head office applications and worldwide FAE team. He is an IEEE member and registered Professional Engineer in Ontario Canada.



## System Level Considerations with GaN Power Switching

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## GaN E-HEMT Characteristics

- Similarities and Differences to MOSFETs

## Low Loss Switching and Conduction in GaN

## GaN E-HEMTs Key Parameters - Reducing System Losses and Parallel Operation

- Gate Charge, On Resistance, Threshold Voltage and Transconductance

## Topologies and Applications Enabled by GaN

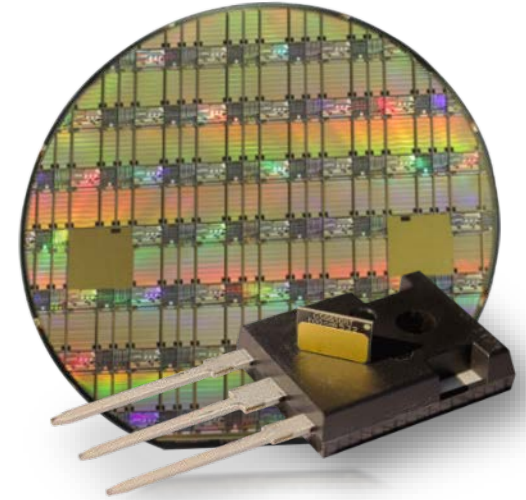
## Thermal Capabilities Providing High Density Power Conversion

## GaN E-HEMT Industry Coming Together to Establish Guidelines & Standards

## Summary

## Market leader for gallium nitride (GaN) power transistors

- Enhancement mode devices
- 100V & 650V devices.
- High current devices (100V to 120A, 650V to 120A)
- Industry's most robust gate drive with +10/-20V range
- GaN<sub>px</sub>® embedded power package for performance



GaN<sub>px</sub>® vs TO-247

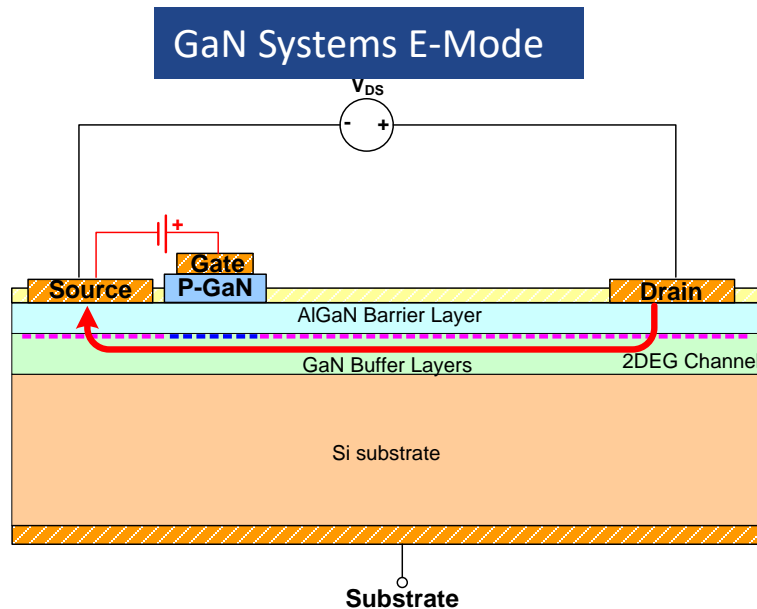
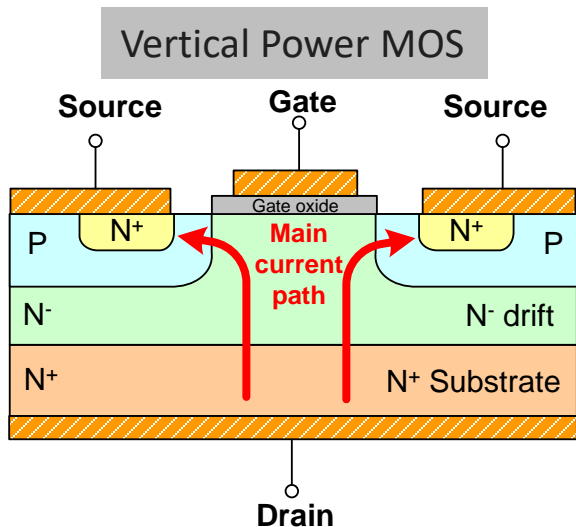
## Global company with decades of GaN experience

- Founded 2008
- Shipping product since 2014
- HQ and R&D in Ottawa, Canada
- Global Sales & Applications support
- World-class fabless manufacturing and advanced packaging



## GaN High Electron Mobility Transistor (HEMT)

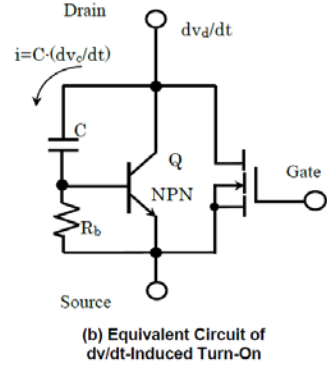
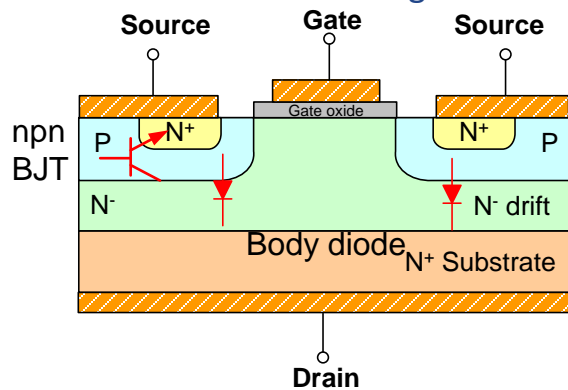
- Epi-growth of GaN on standard high volume silicon wafer,
- 2-Dimensional Electron Gas (2DEG) is formed at the hetero-interface between GaN and AlGaN
- 2DEG has very high charge density and mobility: very low  $R_{DS(on)}$  and high speed device
- P-type GaN depletes 2DEG channel underneath, a positive gate bias turns on the 2DEG – Normally off e-mode
- Voltage driven like Si MOSFET
- No Gate Oxide: high gate reliability



Similar to MOSFET with much faster, more efficient switching

## Si MOSFET

- Parasitic BJT and body diode
- dv/dt Failure
- High reverse recovery charge  $Q_{RR}$
- Not ideal for half bridge hard-switching



### MOSFET dv/dt Limit

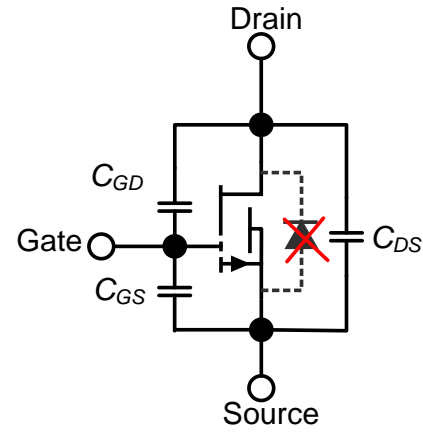
MOSFET dv/dt ruggedness	dv/dt	-	-	80	V/ns
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### MOSFET Body diode $Q_{RR}$

Reverse recovery charge	$Q_{rr}$	-	10	-	$\mu\text{C}$
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## GaN E-HEMT

- No parasitic BJT and body diode
- High dv/dt ruggedness
- Zero reverse recovery
- 2DEG can conduct in reverse, no need for anti-parallel diode
- Fit for half bridge hard & soft switching



GaN E-HEMTs ruggedness, no reverse recovery and reverse conduction provide implementation advantages

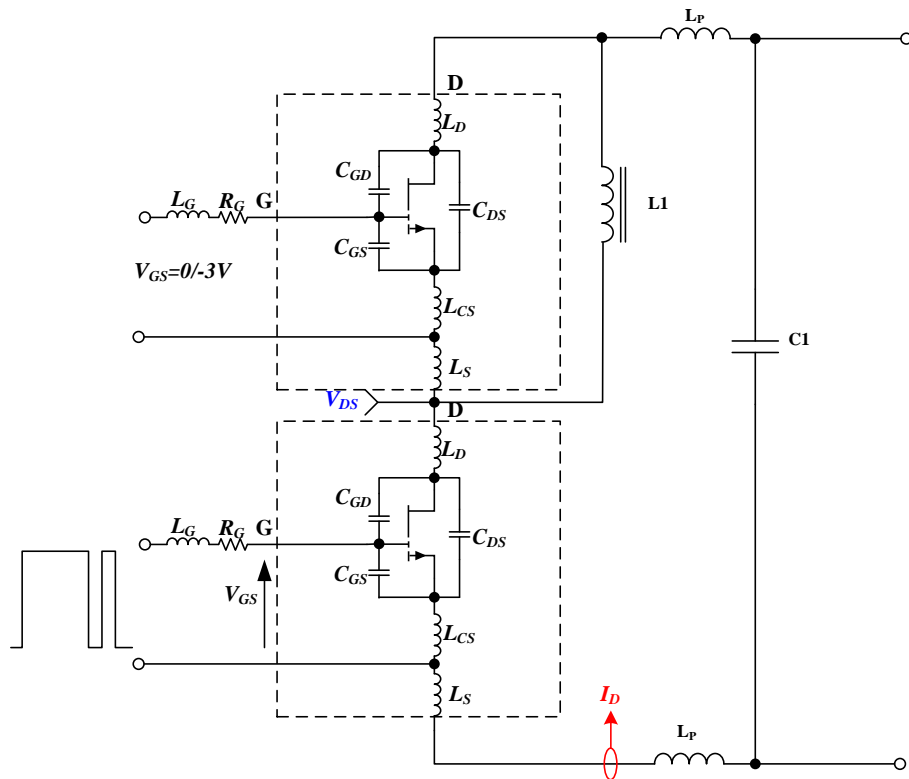


- Circuit model used for explanation
  - $L_G$  – gate loop parasitic inductance
  - $L_D$  – drain parasitic inductance
  - $L_S$  – source parasitic inductance
  - $L_{CS}$  – common parasitic inductance
  - $L_P$  – power loop parasitic inductance
  - $C_{GS}$  = gate to source capacitance
  - $C_{GD}$  – gate to drain capacitance
  - $C_{DS}$  – drain to source capacitance

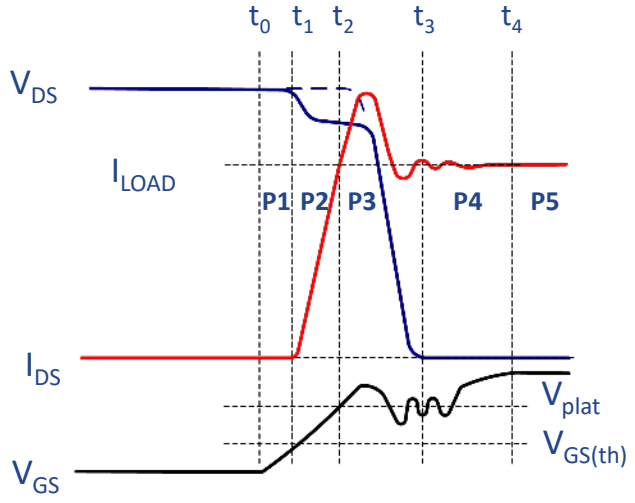
$$C_{ISS} = C_{GS} + C_{GD}$$

$$C_{RSS} = C_{GD}$$

$$C_{OSS} = C_{DS} + C_{GD}$$

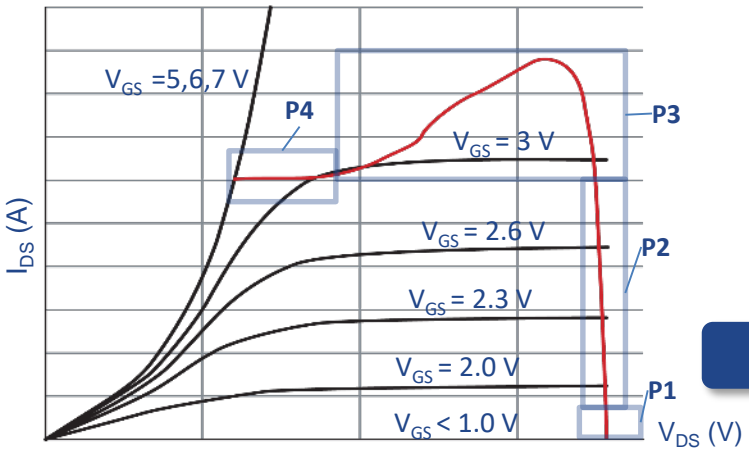


## Consideration of key parameters for GaN



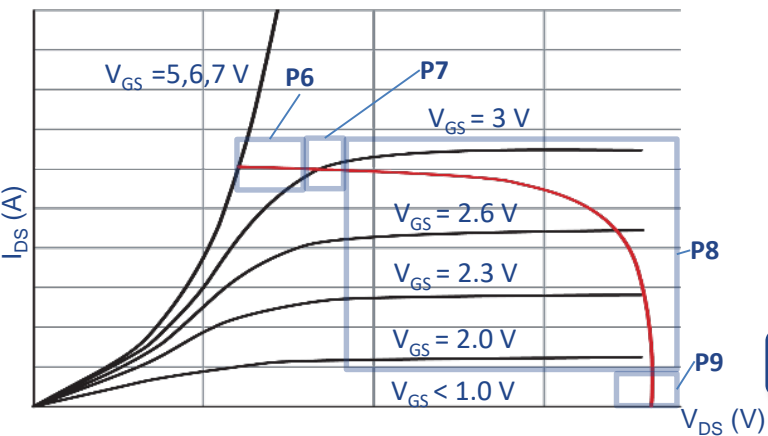
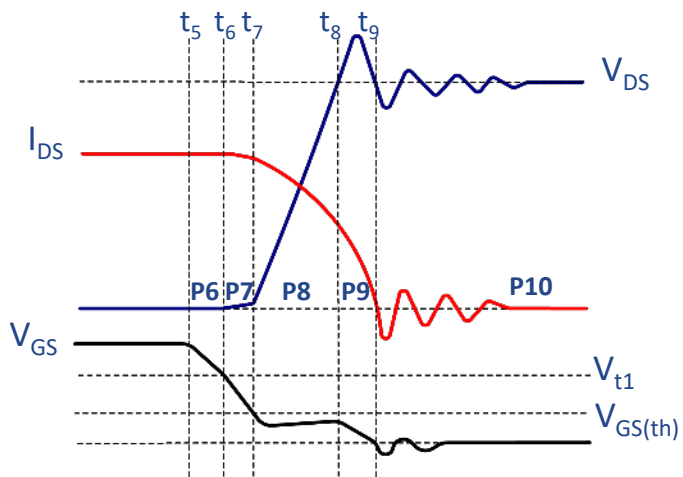
## Low Side Switch Turn On

Period	Time	Summary
P1 ( $t_0 \rightarrow t_1$ )	Delay Period	Small $C_{ISS}$ of GaN E-HEMT results in a low gate driver loss and short delay time.
P2 ( $t_1 \rightarrow t_2$ )	di/dt ( $V \cdot I$ loss) region	Low $C_{ISS}$ and parasitic inductance of GaN PX™ ( $L_G, L_{CS}$ ) allows faster di/dt and lower switching loss with GaN E-HEMT
P3 ( $t_2 \rightarrow t_3$ )	dv/dt (Miller Plateau)	Low $C_{RSS}, C_{OSS}$ and no $Q_{RR}$ are the key to fast turn-on dv/dt and low switching loss of GaN E-HEMT
P4 ( $t_3 \rightarrow t_4$ )	Stabilization	Low inductance of GaN PX™ package reduces the gate overshoot
P5 ( $t_4 \rightarrow t_5$ )	Conduction	Effective $R_{DS(on)} * I_D^2$



An Experimental Comparison of GaN E-HEMTs versus SiC MOSFETs over Different Operating Temperatures by Jianchun Xu, download here: <https://gansystems.com/design-center/papers/>

**Input, Output and Reverse Capacitance Deliver low-loss turn on**



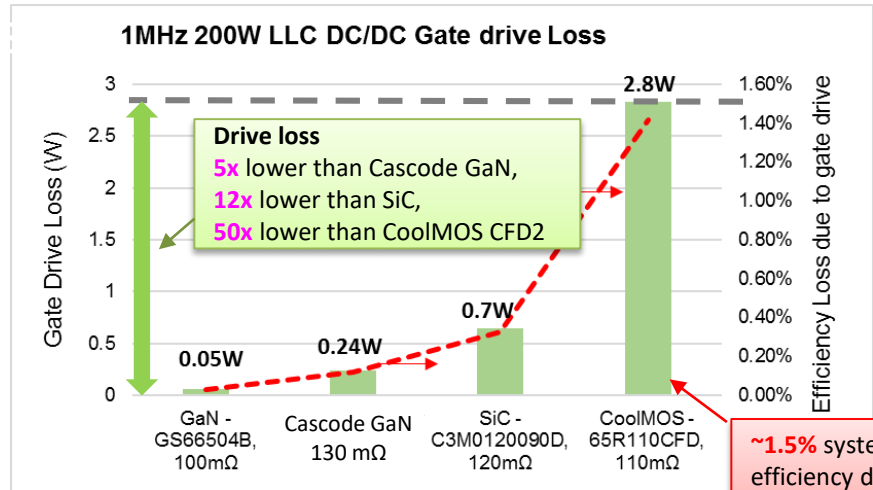
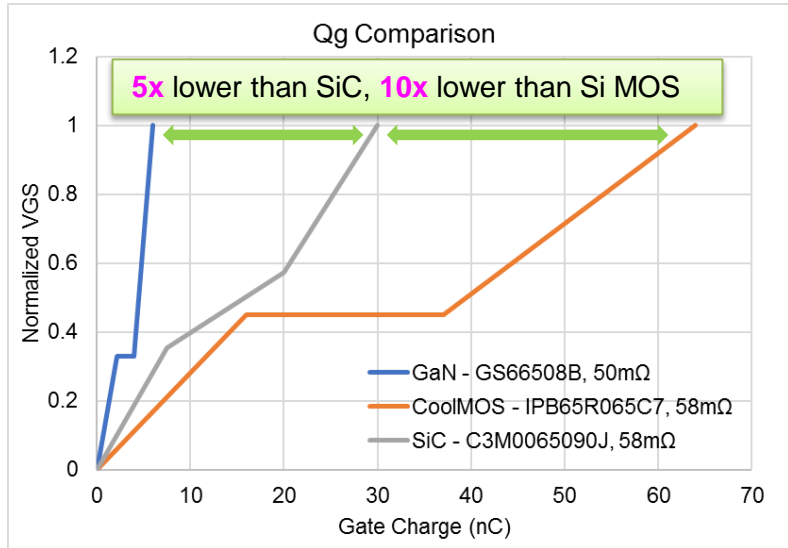
## Low-Side Switch Turn Off

Period	Time	Summary
P6 ( $t_5 \rightarrow t_6$ )	Delay Period	Small $C_{ISS}$ of GaN E-HEMT results in short delay and fast turn-off
P7 ( $t_6 \rightarrow t_7$ )	2DEG Loss Period	Due to the ultra-fast transition of GaN E-HEMT, switching loss at turn-off is very small
P8 ( $t_7 \rightarrow t_8$ )	$dv/dt$ ( $C_{OSS}$ charging)	Fast transition of GaN E-HEMT causes 2DEG to open as $V_{DS}$ rises resulting in no loss in $dv/dt$ (Miller plateau) period
P9 ( $t_8 \rightarrow t_9$ )	$di/dt$ (overshoot)	Low parasitic inductance and Kelvin source sense of GaNPX™ package minimizes the spike and ringing.
P10 ( $t_9 \rightarrow t_1$ )	Off time	

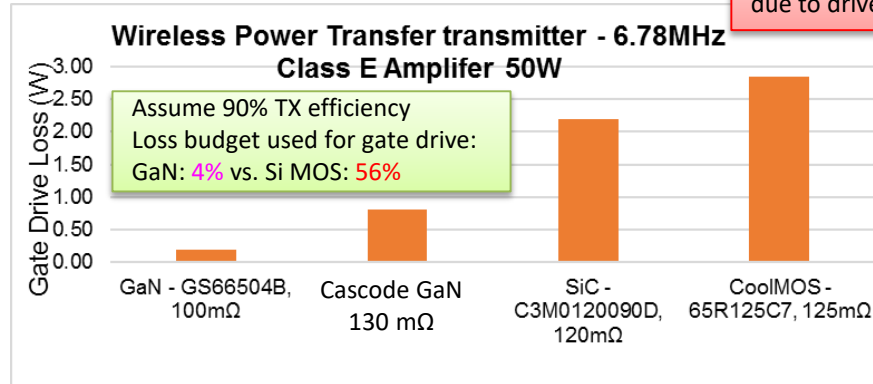
An Experimental Comparison of GaN E-HEMTs versus SiC MOSFETs over Different Operating Temperatures by Jianchun Xu, download here: <https://gansystems.com/design-center/papers/>

Turn off switching loss greatly reduced compared to MOSFET

- 5-10x smaller than equivalent Si MOSFET
- Up to 50x lower gate drive loss at MHz switching



~1.5% system efficiency drop due to drive loss



Low  $Q_G$  key to efficient high  $f_{sw}$  operation

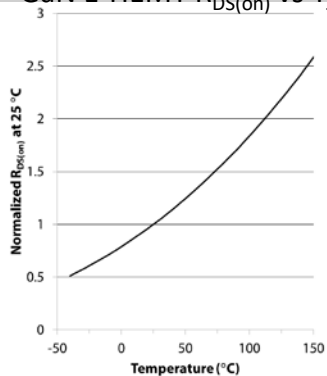
## On Resistance Temperature Dependency

- GaN E-HEMT has positive temperature coefficient  $R_{DS(on)}$
- Helps the current sharing in parallel operation

## Threshold Voltage Temperature Dependency

- GaN E-HEMT has stable gate threshold over the temperature range
- Si/SiC MOSFET  $V_{GS(th)}$  decreases with temperature:
  - Hotter drive turn-on earlier potential runaway

GaN E-HEMT  $R_{DS(on)}$  vs  $T_J$



SiC  $R_{DS(on)}$  vs  $T_J$

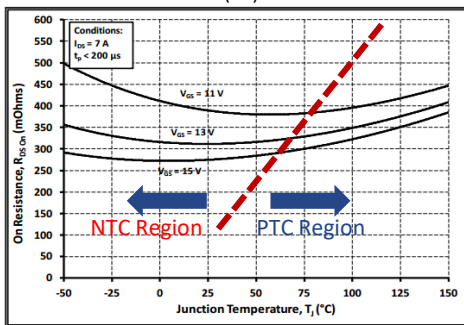
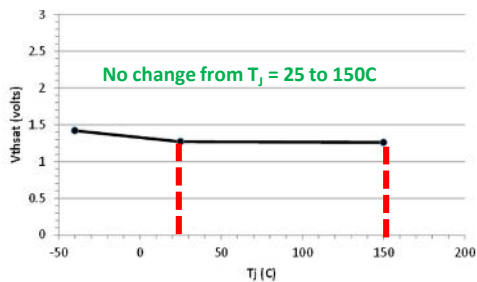
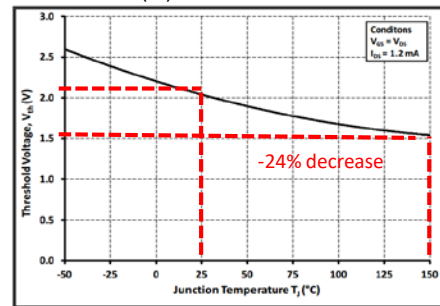


Figure 6. On-Resistance vs. Temperature For Various Gate Voltage

GaN E-HEMT  $V_{GS(th)}$  over  $T_J$  range



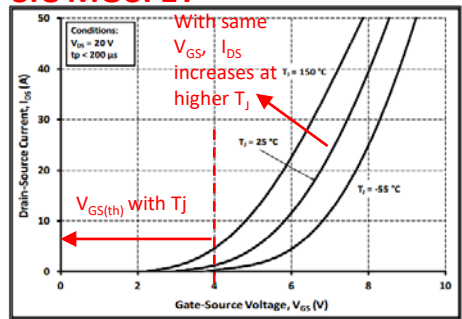
SiC  $V_{GS(th)}$  decreases with  $T_J$



Positive TC for  $R_{DS(on)}$  and stable  $V_{GS(th)}$  eliminates current mismatch and thermal runaway

- GaN E-HEMT Trans-conductance  $g_m$  decreases with temperature, good for paralleling

## SiC MOSFET



SiC:  $V_{GS(th)}$  drops and  $g_m$  increases:

- Hotter device has higher switching current -> higher switching loss
- Positive feedback, potential thermal runaway if not designed properly

## GaN HEMT

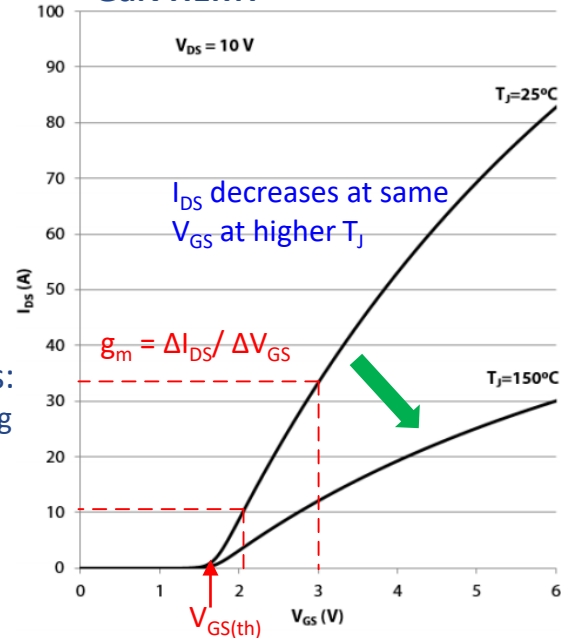
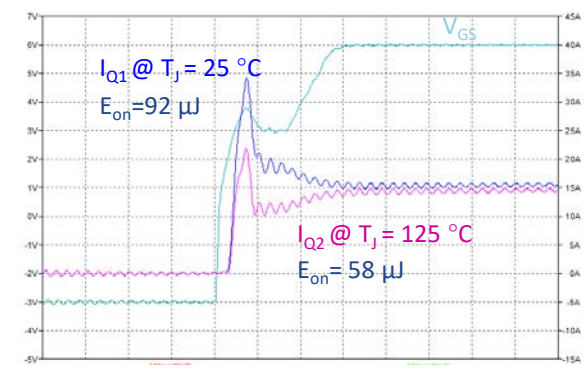


Figure 10: Typical  $I_{DS}$  vs.  $V_{GS}$

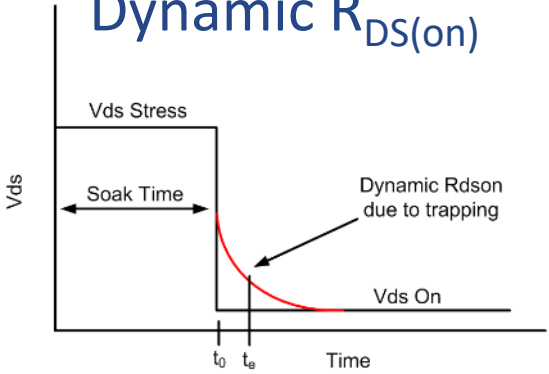
- Negative feedback for self balancing in parallel:
- $T_j \uparrow - g_m \downarrow - I_{D@switching} \downarrow - E_{on} \downarrow - T_j \downarrow$

2x GS66508T paralleled 400V/30A turn-on waveforms with different  $T_j$



Dynamic Current Sharing and Self Balancing achieved with GaN  $g_M$  and  $V_{GS(th)}$  characteristics

## Dynamic $R_{DS(on)}$



## Factors Affecting Dyn $R_{DS(on)}$

- Bias Voltage
- Bias time
- Junction Temperature
- Switching Frequency
- Duty Cycle

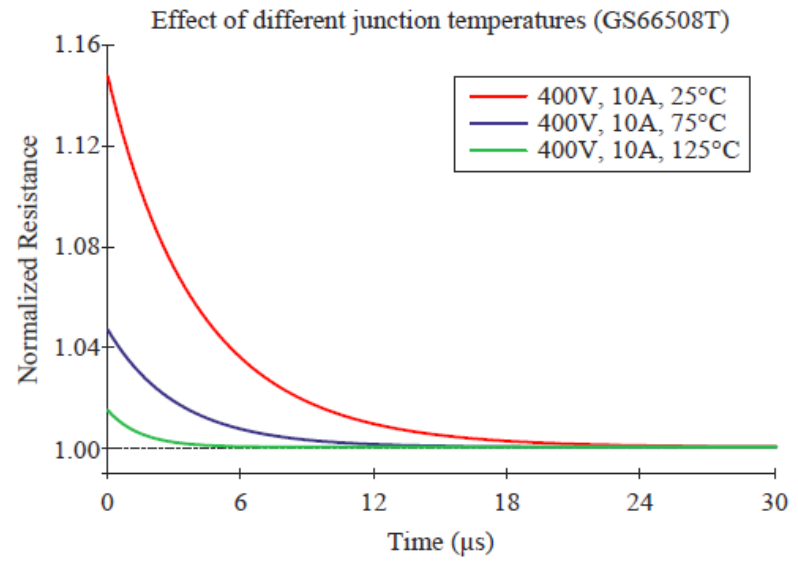


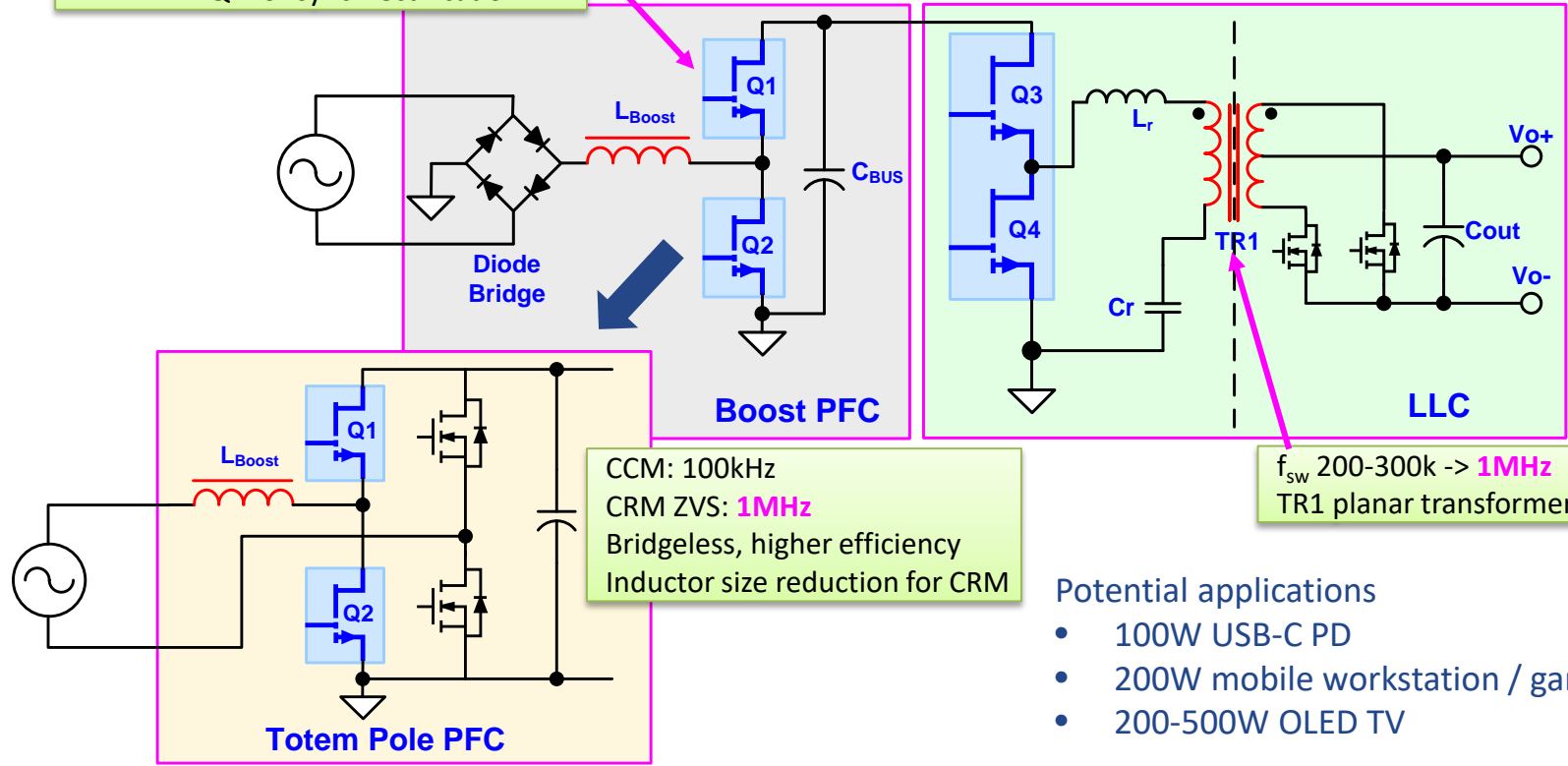
Fig. 14. Normalized  $R_{dyn}$  of GS66508T for different junction temperatures

Oliveira, E., Nöding, C., Zacharias, P., (May, 2017) Impact of Dynamic On-Resistance of High Voltage GaN Switches on the Overall Conduction Losses, ISBN 978-3-8007-4424-4

Dynamic  $R_{DS(on)}$  affects are negligible in steady state operation

- $f_{sw}$  100k -> **200kHz**
- Reduces  $L_{boost}$  size
- D1 -> Q1 for sync. rectification

## >75W AC/DC: PFC+LLC



CCM: 100kHz  
 CRM ZVS: **1MHz**  
 Bridgeless, higher efficiency  
 Inductor size reduction for CRM

$f_{sw}$  200-300k -> **1MHz**  
 TR1 planar transformer reduce height

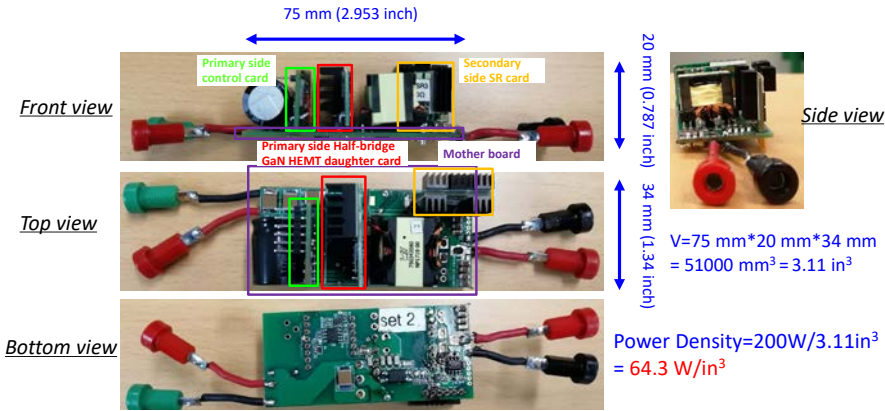
### Potential applications

- 100W USB-C PD
- 200W mobile workstation / gaming laptop
- 200-500W OLED TV

**GaN enhances PFC and LLC and enables BTP PFC**

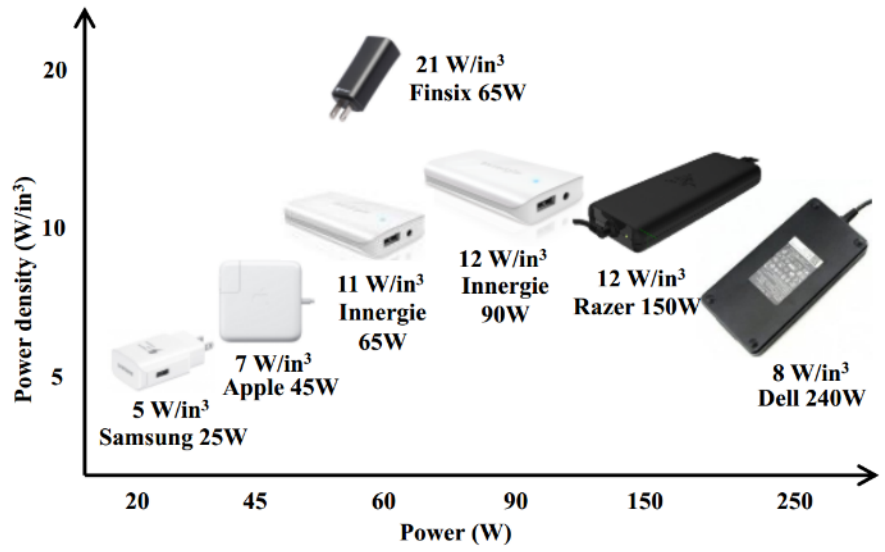
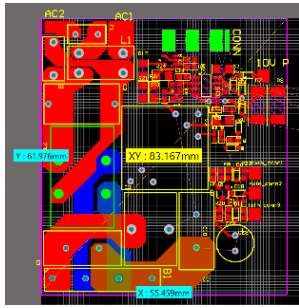


## LLC DC/DC stage prototype



- 2-3x smaller size than existing adapter on market
- Efficiency target: 93% total, 96% LLC
- Overall power density (target): **27.36 W/in<sup>3</sup>**

## PFC stage board design (4.2 in<sup>3</sup>)



Achieves Highest Power Density

## The Value of GaN

- No parasitic BJT and Body Diode
- **Zero** Reverse Recovery
- Reverse conduction, anti-Parallel Diode not required

## Smaller Size

- Higher frequency → Smaller filters
- Better topology → Fewer parts
- Lower loss → Smaller heatsink

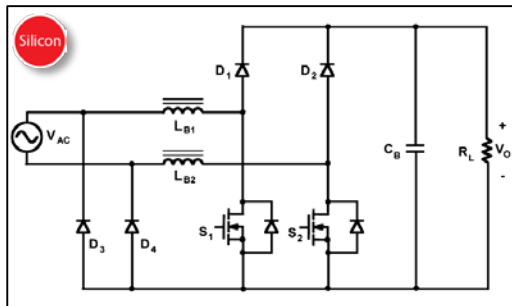
## Better Efficiency

- Increases from 97.5% to 99%

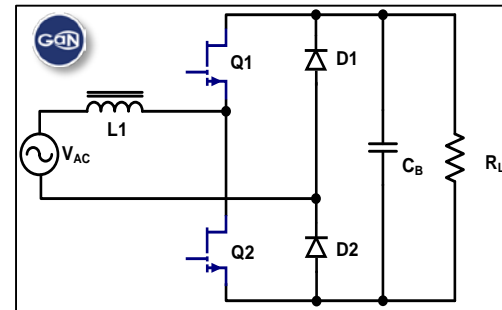
## Lower System Cost

- Smaller EMI filter → less expensive
- Smaller heatsink → less expensive

### Semi-bridgeless PFC – Si MOSFET

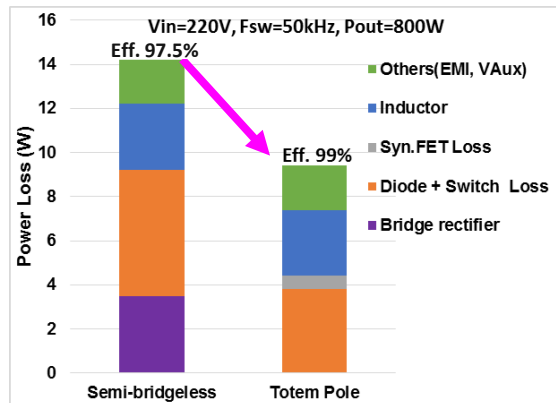


### Totem Pole PFC - GaN

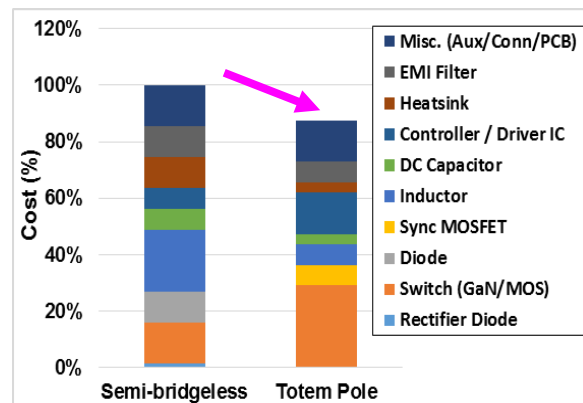


Part count reduced by 33%

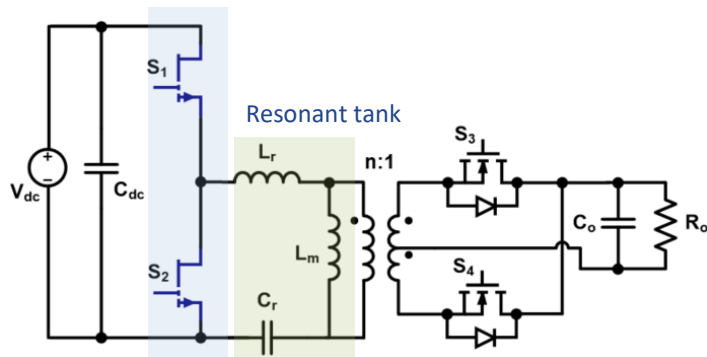
### PFC Loss Breakdown → 15% lower



### BOM Cost Improvement → 15% lower



GaN makes smaller, more efficient, lower cost PFC solutions possible

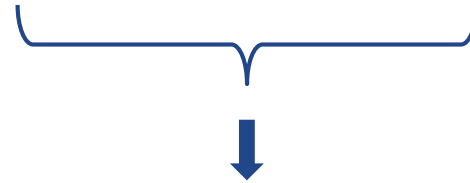


GaN-based half-bridge LLC converter

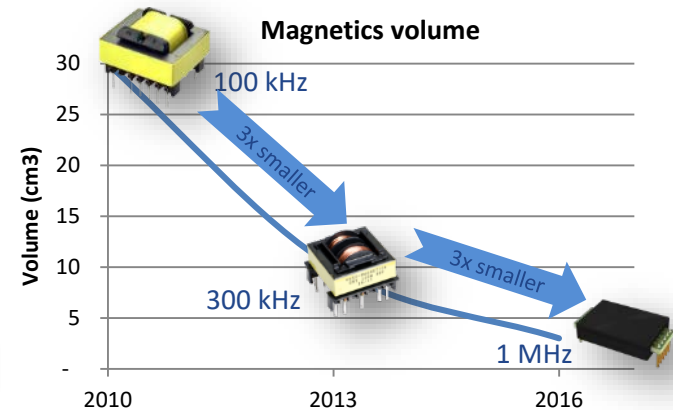
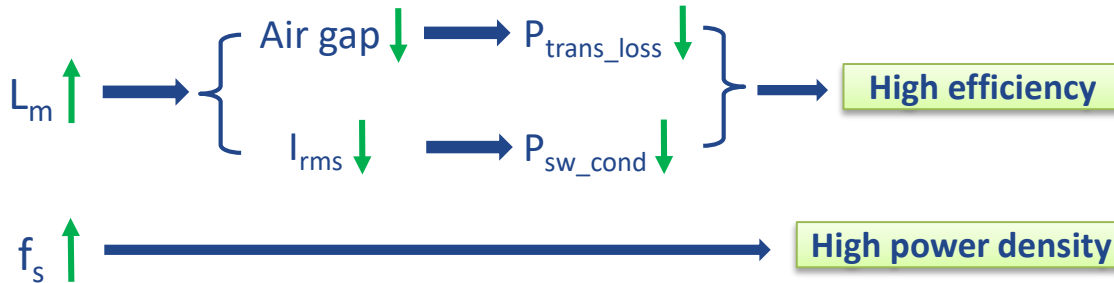
## The ZVS constraints of LLC converter

Sufficient  
deadtime

Sufficient inductive  
energy from  $L_m$

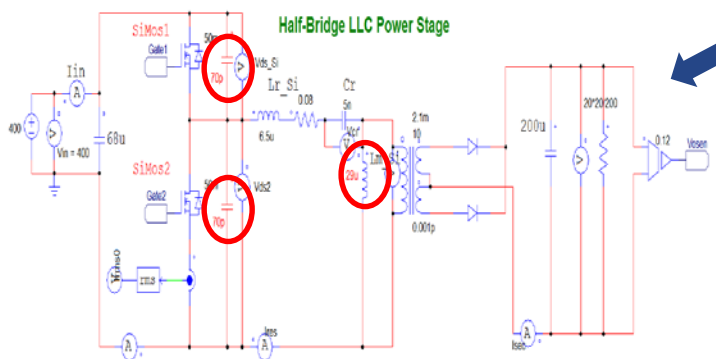


A smaller  $C_{oss}$  leads to:

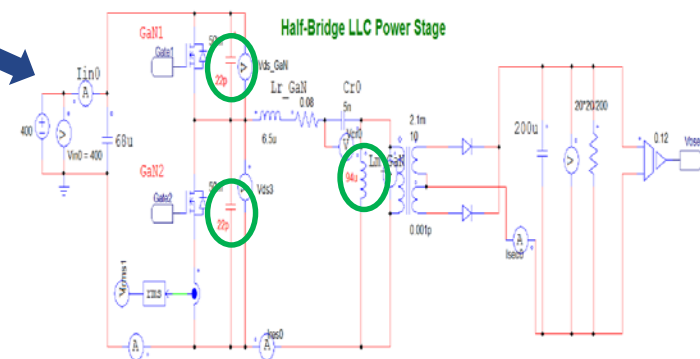


Better switching characteristics of GaN improve LLC efficiency and size

## Simulation comparison between Si-based and GaN-based LLC



Converter design specification	
$V_{in}$	300-400 V
$V_o$	20 V
$I_o$	10 A
$P_o$	200 W
$n$	10:1:1
$F_s$	900 kHz
$T_{dead}$	60 ns

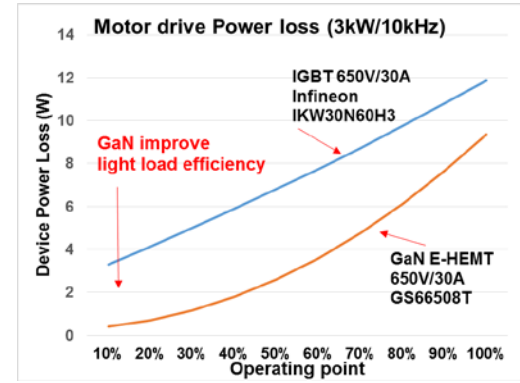
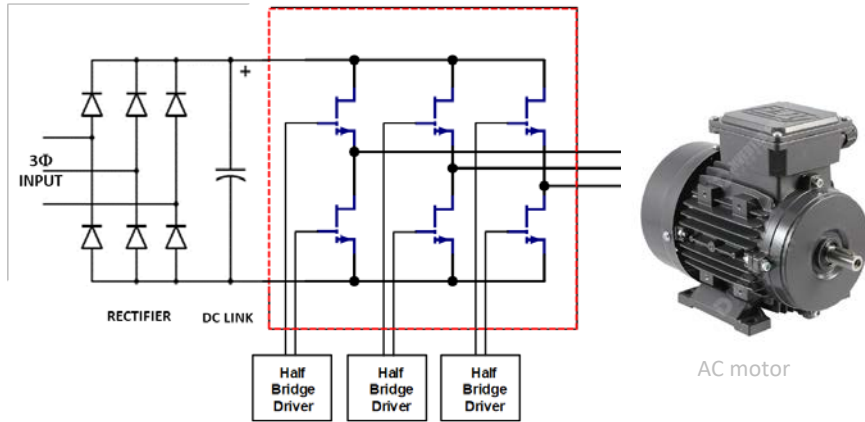


- Same converter requirement is applied to both solutions
- 68% increase on the magnetizing inductance can be achieved with GaN version

Switch parameter comparison	Si MOSFET	GaN HEMT	Unit
	IPA65R190CFD	GS66502B	
$V_{DS}$	650	650	V
$R_{DS(ON)}$	190	200	m $\Omega$
$Q_G$	68	1.5	nC
$C_{O(ER)}$	70	22	pF
$C_{O(TR)}$	336	35	pF

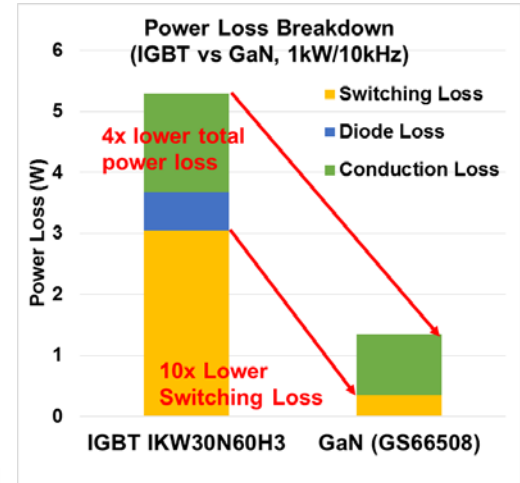
Design parameter comparison	Si MOSFET	GaN HEMT	Unit
	IPA65R190CFD	GS66502B	
$L_{m\_max}$	29	94	$\mu$ H
$I_{rms}$	1.3	0.8	A
$I_{pk}$	2.7	1.9	A
$P_{cond}$	0.32	0.13	W

GaN based LLC: 60% lower condition losses, 75X lower gate drive loss



## The Value of GaN

- **High  $f_{sw}$** 
  - No acoustic noise for quiet operation
  - Smaller and less expensive filtering
  - Sinusoidal output filter for less expensive unshielded cabling, and longer cable lengths
- **Lower Losses**
  - Smaller less expensive heatsinks
  - High Efficiency to meet stringent regulations



GaN Provides Low Frequency, High Current Efficiency and Size

*GSP65R25HB-EVB or GSP65R13HB-EVB*

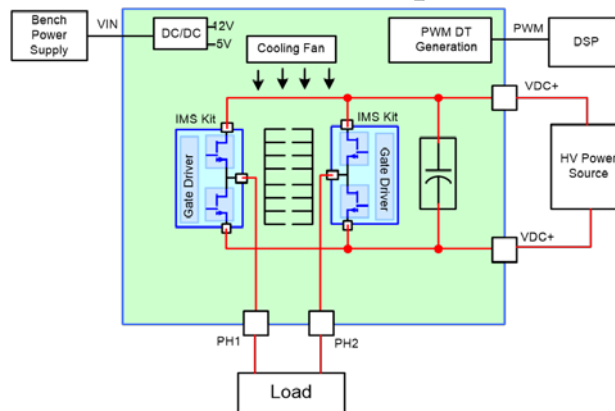
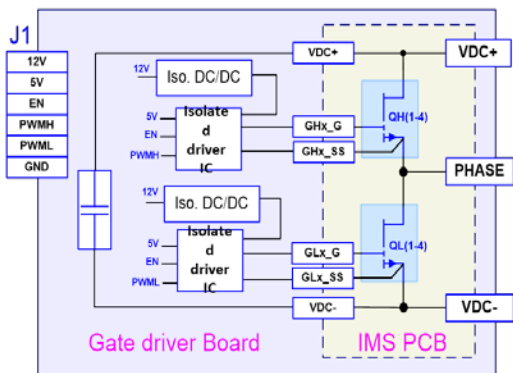
*GSP65MB-EVB*



GaN High Power Half Bridge evaluation assembly (Gate Driver pcb + IMS HB)

IMS Half Bridge Boards (wo parallel and single transistor)

High Power Full Bridge evaluation platform mother board



Uses design concept of higher power GaN intelligent power modules (IPM)

# GaNSPEC - GaN Standards for Power Electronic Conversion Devices Working Group

## Mission Statement

We seek to create standards and guidelines for Test Methods, Reliability & Qualification Procedures, and Datasheet Parametrics for GaN based power conversion devices



*GaN Systems works with industry to standardize GaN information and is founding member of the new JEDEC committee: JC-70 Wide Bandgap Power Electronic Conversion Semiconductors*

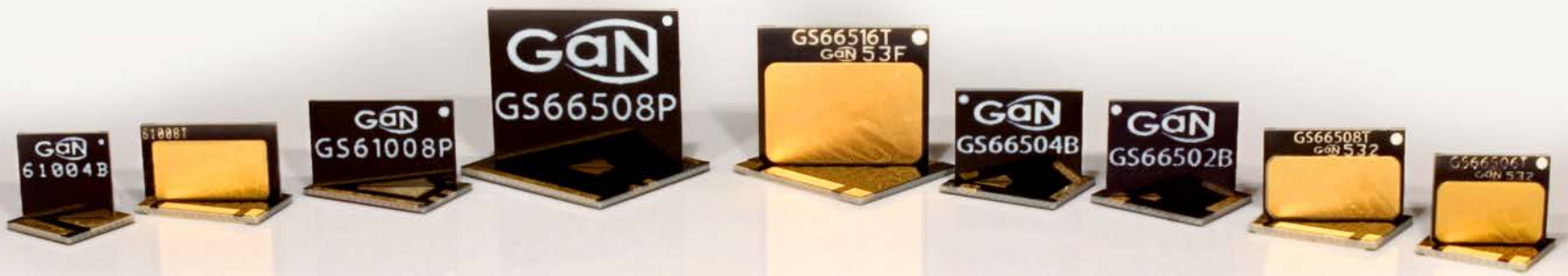
## Standards for Wide Bandgap Device Qualification, Test Methods and Datasheet Parameters

- Members from more than 30 companies: power semiconductor suppliers, users, supply chain representatives.
- Both GaN (JC-70.1) and SiC (JC-70.2) device subcommittees formed
- Interested companies worldwide are welcome to join JEDEC to participate in this important standardization effort.
- Find more information about membership
  - <https://www.jedec.org/join-jedec>
- Contact [Emily Desjardins](mailto:emilyd@jedec.org) to learn more ([emilyd@jedec.org](mailto:emilyd@jedec.org))
- Get informed! At APEC2018: attend a presentation on JC-70 activity at Industry Session IS16, Thursday March 8 from 8:30 to 11:55.



- GaN E-HEMTs are mainstream in power electronics design
- Enable topologies and applications
- Reduced losses, weight and volume
- Easy to Use, More Combined Knowledge, Fewer Unknowns
- Lower costs: system, shipping, installation, maintenance and operating costs

Thank You. Please Visit GaN Systems at booth # 1041 for more information and discussion



Questions?

